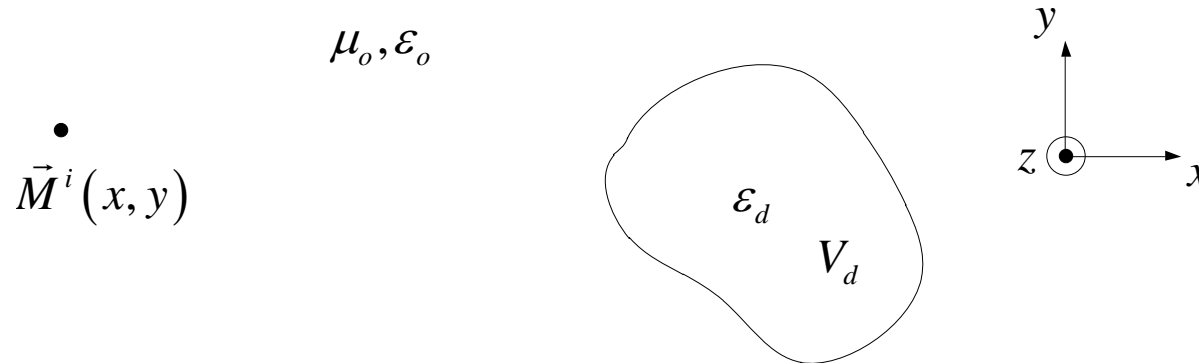
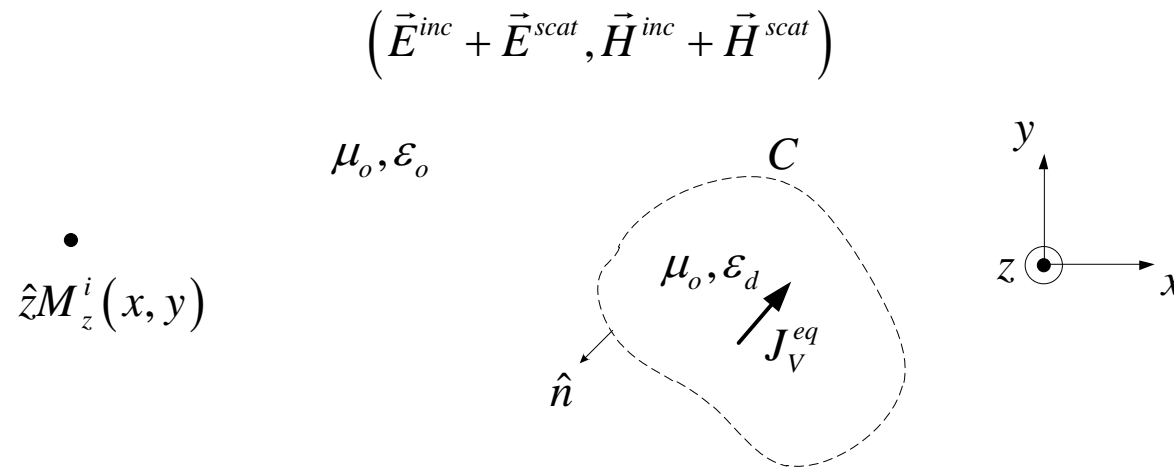


TE_z-Scattering by a Dielectric Cylinder – VEFIE Solution



- Consider a magnetic current density $\vec{M}^i(x, y) = \hat{z}M_z^i$ that is infinitely long and invariant w.r.t. the z -axis radiating in the presence of an infinitely long dielectric cylinder defined by a volume V_d that is situated in an infinite, homogeneous, unbounded media.
- Objective:
 - Compute the total fields inside and outside of the dielectric cylinder
- Solution
 - Pose the equivalent problem using volume equivalence
 - Derive the Volume Electric Field Integral Equation for TE-z polarization
 - Solve for the equivalent currents via the Method of Weighted Residuals
 - Compute the scattered fields from the equivalent currents

The Equivalent Problem



- The dielectric is replaced with an electric volume equivalent current density that is radiating in a homogeneous free space:

$$\vec{J}_V^{eq}(\vec{r}) = j\omega\epsilon_o(\epsilon_r - 1)\vec{E}^{tot}(\vec{r})$$

where $\epsilon_d = \epsilon_r\epsilon_o$

- Note that ϵ_r can be complex (lossy material), and can be inhomogeneous.
- We derived the VEFIE by enforcing $\vec{E}^{inc}(\vec{r}) = \vec{E}^{tot}(\vec{r}) - \vec{E}^{scat}(\vec{r})$, $\vec{r} \in V_d$:

$$\vec{E}^{inc}(\vec{r}) = \frac{\vec{J}_V^{eq}(\vec{r})}{j\omega\epsilon_o(\epsilon_r - 1)} - \eta_o \frac{\nabla\nabla \cdot \vec{A}(\vec{r}) + k_o^2\vec{A}(\vec{r})}{jk_o}$$

- where

$$\vec{A}(\vec{r}) = \int_{S_d} \frac{1}{4j} H_0^{(2)}(k_o|\vec{r} - \vec{r}'|) \vec{J}_V^{eq}(\vec{r}') ds' = \hat{x}A_x(\vec{r}) + \hat{y}A_y(\vec{r})$$

- The $\nabla\nabla \cdot \vec{A}(\vec{r})$ can be expressed as:

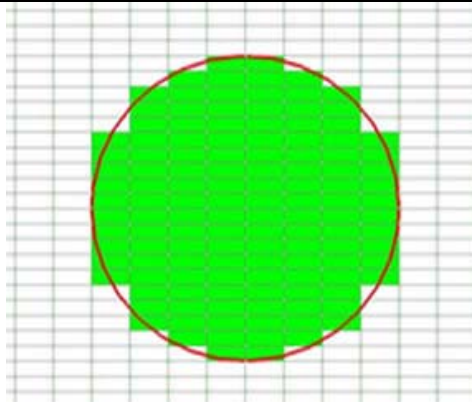
$$\nabla\nabla \cdot \vec{A}(\vec{r}) = \hat{x} \left(\frac{\partial^2 A_x(\vec{r})}{\partial x^2} + \frac{\partial^2 A_y(\vec{r})}{\partial x \partial y} \right) + \hat{y} \left(\frac{\partial^2 A_y(\vec{r})}{\partial y^2} + \frac{\partial^2 A_x(\vec{r})}{\partial y \partial x} \right)$$

- Then, breaking up the VEFIE into vector projections, it can be expressed as:

$$j \frac{k_o}{\eta_o} \begin{bmatrix} E_x^{inc}(\vec{r}) \\ E_y^{inc}(\vec{r}) \end{bmatrix} = \frac{1}{\epsilon_r - 1} \begin{bmatrix} 1 & 0 \\ 0 & 1 \end{bmatrix} \begin{bmatrix} J_{V_x}^{eq}(\vec{r}) \\ J_{V_y}^{eq}(\vec{r}) \end{bmatrix} - \begin{bmatrix} k_o^2 + \frac{\partial^2}{\partial x^2} & \frac{\partial^2}{\partial x \partial y} \\ \frac{\partial^2}{\partial x \partial y} & k_o^2 + \frac{\partial^2}{\partial y^2} \end{bmatrix} \begin{bmatrix} A_x(\vec{r}) \\ A_y(\vec{r}) \end{bmatrix}$$

- Next, the MoM solution of the TE_z-VEFIE will be pursued. To this end, $J_{V_x}^{eq}$ and $J_{V_y}^{eq}$ will be treated as unknowns.

The Method of Moment Solution



- Mesh
 - Rectangular cell approximation of the object
 - Note, this is a low-order approximation (simplest)
 - Triangles or arbitrary quads are more accurate
 - Curvilinear cells are most accurate
- Current basis function space:
 - Pulse approximation (piece-wise constant)
 - $J_{V_x}^{eq} \approx \sum_{n=1}^N \alpha_{x,n} P_n(\vec{r})$, $J_{V_y}^{eq} \approx \sum_{n=1}^N \alpha_{y,n} P_n(\vec{r})$, $P_n(\vec{r}) = \begin{cases} 1, & \vec{r} \in S_n \\ 0, & \text{else} \end{cases}$
- Test function space:
 - Point-matching at the cell centers
 - $T_{x_m} = \delta(|\vec{r} - \vec{r}_m^c|)$, $T_{y_m} = \delta(|\vec{r} - \vec{r}_m^c|)$

Staircase-Approximation



Computing the Vector Potential

- Define

$$\circ A_{x,y}(\vec{r}) = \frac{1}{4j} \iint_S J_{V_x,y}^{eq}(\vec{r}) H_0^{(2)}(k_o R) ds' \approx \frac{1}{4j} \sum_{n=1}^N \alpha_{x,y,n} \iint_{S_n} H_0^{(2)}(k_o R) ds'$$

- Again, use Richmond's approximation by approximating each rectangular cell as a circle with radius:

$$\blacksquare a_n = \sqrt{\frac{\ell_n h_n}{\pi}}$$

- where ℓ_n and h_n are the dimensions of the cell

- The vector potential can be approximated as:

$$\circ A_{x,y}(\vec{r}) \approx \sum_{n=1}^N \alpha_{x,y,n} \frac{1}{4j} \int_0^{a_n} \int_0^{2\pi} H_0^{(2)}(k_o R) \rho' d\phi' d\rho'$$

- where $R = |\vec{r} - \vec{r}'|$

- To compute the integration, use the addition theorem:

$$\circ H_0^{(2)}(k_o|\vec{r}-\vec{r}'|) = \begin{cases} \sum_{\nu=-\infty}^{\infty} J_\nu(k_o\rho')H_\nu^{(2)}(k_o\rho)e^{j\nu(\phi-\phi')}, & \rho > \rho' \\ \sum_{\nu=-\infty}^{\infty} J_\nu(k_o\rho)H_\nu^{(2)}(k_o\rho')e^{j\nu(\phi-\phi')}, & \rho < \rho' \end{cases}$$

- where (ρ, ϕ) are the coordinates of \vec{r} , and (ρ', ϕ') of \vec{r}'

- For non-self-terms, the field point will be outside the source cell. Thus $\rho > \rho'$:

$$\circ \frac{1}{4j} \int_0^{a_n} \int_0^{2\pi} H_0^{(2)}(k_o R) \rho' d\phi' d\rho' = \frac{1}{4j} \int_0^{a_n} \int_0^{2\pi} \sum_{\nu=-\infty}^{\infty} J_\nu(k_o\rho')H_\nu^{(2)}(k_o\rho_n)e^{j\nu(\phi_m-\phi')} \rho' d\phi' d\rho'$$

- where ρ_n is the radial distance from \vec{r}_n^c ($\rho_n = \sqrt{(x-x_n)^2 + (y-y_n)^2}$)

- Note, that the ϕ -integration is only non-zero when $\nu = 0$

$$\circ \frac{2\pi}{4j} H_0^{(2)}(k_o\rho_n) \int_0^{a_n} J_0(k_o\rho') \rho' d\phi' d\rho' = \frac{\pi}{2j} H_0^{(2)}(k_o\rho_n) \left(\frac{\rho'}{k_o} J_1(k_o\rho') \Big|_0^{a_n} \right)$$

$$\circ = \frac{\pi a_n}{2jk_o} H_0^{(2)}(k_o\rho_n) J_1(k_o a_n)$$

- Therefore,

$$\circ A_{x,n}(\vec{r}) \approx \alpha_{x,n} \frac{\pi a_n}{2jk_o} J_1(k_o a_n) H_0^{(2)}(k_o\rho_n), \text{ for } \vec{r} \notin S_n$$

- Recall, the VEFIE required derivatives of A in the observation coordinates

$$\circ \left(k_o^2 + \frac{\partial^2}{\partial x^2} \right) A_{x,n}(x, y) \approx \alpha_{x,n} K_n \left(k_o^2 + \frac{\partial^2}{\partial x^2} \right) H_0^{(2)}(k_o \rho_n)$$

$$\blacksquare \text{ where: } K_n = \frac{\pi a_n}{2jk_o} J_1(k_o a_n)$$

- Computing:

$$\circ \frac{\partial}{\partial x} H_0^{(2)}(k_o \rho_n) = k_o \frac{\partial}{\partial x} \rho_n H_0^{(2)'}(k_o \rho_n) = -k_o \frac{x - x_n}{\rho_n} H_1^{(2)}(k_o \rho_n)$$

$$\circ \frac{\partial^2}{\partial x^2} H_0^{(2)}(k_o \rho_n) = \frac{\partial}{\partial x} \left(-k_o \frac{x - x_n}{\rho_n} H_1^{(2)}(k_o \rho_n) \right)$$

$$= -\frac{k_o}{\rho_n} H_1^{(2)}(k_o \rho_n) - k_o^2 \left(\frac{x - x_n}{\rho_n} \right)^2 H_0^{(2)}(k_o \rho_n) + 2k_o \frac{(x - x_n)^2}{\rho_n^3} H_1^{(2)}(k_o \rho_n)$$

- Noting that:

$$-\frac{k_o}{\rho_n} H_1^{(2)}(k_o \rho_n) + 2k_o \frac{(x - x_n)^2}{\rho_n^3} H_1^{(2)}(k_o \rho_n) = -k_o \frac{(\rho_n^2 - 2(x - x_n)^2)}{\rho_n^3} H_1^{(2)}(k_o \rho_n)$$

$$= -k_o \frac{((y - y_n)^2 - (x - x_n)^2)}{\rho_n^3} H_1^{(2)}(k_o \rho_n)$$

○ Also, noting that

$$\blacksquare k_o^2 H_0^{(2)}(k_o \rho_n) - k_o^2 \frac{(x - x_n)^2}{\rho_n^2} H_0^{(2)}(k_o \rho_n) = k_o^2 \frac{(y - y_n)^2}{\rho_n^2} H_0^{(2)}(k_o \rho_n)$$

• Combining these, finally leads to:

$$\left(k_o^2 + \frac{\partial^2}{\partial x^2} \right) H_0^{(2)}(k_o \rho_n) = k_o^2 \frac{(y - y_n)^2}{\rho_n^2} H_0^{(2)}(k_o \rho_n) + k_o \frac{(x - x_n)^2 - (y - y_n)^2}{\rho_n^3} H_1^{(2)}(k_o \rho_n) = F_{x,x}(x, y)$$

• By duality:

$$\left(k_o^2 + \frac{\partial^2}{\partial y^2} \right) H_0^{(2)}(k_o \rho_n) = k_o^2 \frac{(x - x_n)^2}{\rho_n^2} H_0^{(2)}(k_o \rho_n) + k_o \frac{(y - y_n)^2 - (x - x_n)^2}{\rho_n^3} H_1^{(2)}(k_o \rho_n) = F_{y,y}(x, y)$$

• It can also be shown that:

$$\frac{\partial^2}{\partial x \partial y} H_0^{(2)}(k_o \rho_n) = k_o \frac{(y - y_n)(x - x_n)}{\rho_n^3} \left[2H_1^{(2)}(k_o \rho_n) - k_o \rho_n H_0^{(2)}(k_o \rho_n) \right] = F_{x,y}(x, y)$$

○ where

$$\blacksquare \rho_n = \sqrt{(x - x_n)^2 + (y - y_n)^2}$$

The Impedance Matrix

- Applying point matching, the discrete VEFIE can be expressed as:

$$j \frac{k_o}{\eta_o} \begin{bmatrix} E_x^{inc}(\vec{r}_m) \\ E_y^{inc}(\vec{r}_m) \end{bmatrix} = \begin{bmatrix} [A_{m,n}] & [B_{m,n}] \\ [C_{m,n}] & [D_{m,n}] \end{bmatrix} \begin{bmatrix} \bar{\alpha}_x \\ \bar{\alpha}_y \end{bmatrix}$$

○ where, $[A_{m,n}]$, etc., are matrix blocks.

- For the non-self terms ($m \neq n$),

$$A_{m,n} = -K_n F_{x,x}(x_m, y_m), \quad B_{m,n} = -K_n F_{x,y}(x_m, y_m),$$

$$C_{m,n} = -K_n F_{x,y}(x_m, y_m), \quad D_{m,n} = -K_n F_{y,y}(x_m, y_m).$$

- For example:

$$\circ A_{m,n} = -K_n \left[k_o^2 \frac{(y_m - y_n)^2}{\rho_{m,n}^2} H_0^{(2)}(k_o \rho_{m,n}) + k_o \frac{(x_m - x_n)^2 - (y_m - y_n)^2}{\rho_{m,n}^3} H_1^{(2)}(k_o \rho_{m,n}) \right]$$

○ where

$$\blacksquare \rho_{m,n} = \sqrt{(x_m - x_n)^2 + (y_m - y_n)^2}$$

Evaluating the Self Term

- Richmond's method is again used for the self-term:

$$\circ \frac{1}{4j} \iint_{S_n} H_0^{(2)}(k_o R_n) ds' \approx \frac{1}{4j} \int_0^{a_n} \int_0^{2\pi} H_0^{(2)}(k_o R_n) \rho' d\phi' d\rho'$$

- Next, applying the addition theorem for $\rho < \rho'$

$$\circ \frac{1}{4j} \int_0^{a_n} \int_0^{2\pi} H_0^{(2)}(k_o R_n) \rho' d\phi' d\rho' = \frac{1}{4j} \left(\frac{2\pi a_n}{k_o} J_0(k_o \rho_n) H_1^{(2)}(k_o a_n) - j \frac{4}{k_o^2} \right)$$

- Note that ρ_n is the observation coordinate (distance from the center)

- Differentiating in the observation coordinates:

$$\circ \left(k_o^2 + \frac{\partial^2}{\partial x^2} \right) J_0^{(2)}(k_o \rho_n) = k_o^2 \frac{(y - y_n)^2}{\rho_n^2} J_0(k_o \rho_n) + k_o \frac{(x - x_n)^2 - (y - y_n)^2}{\rho_n^3} J_1(k_o \rho_n)$$

- For the self-term, this must be evaluated in the limit that $(x, y) \rightarrow (x_n, y_n)$

- Note:

$$\blacksquare \frac{(y - y_n)}{\rho_n} = \sin \phi_n, \quad \frac{(x - x_n)}{\rho_n} = \cos \phi_n$$

- Therefore:

$$\circ \left(k_o^2 + \frac{\partial^2}{\partial x^2} \right) J_0^{(2)}(k_o \rho_n) = k_o^2 \sin^2 \phi_n J_0(k_o \rho_n) + k_o \frac{(\cos \phi_n)^2 - (\sin \phi_n)^2}{\rho_n} J_1(k_o \rho_n)$$

- Now, take the limit as $\rho_n \rightarrow 0$

$$\begin{aligned} \lim_{\rho_n \rightarrow 0} \left[k_o^2 \sin^2 \phi_n J_0(k_o \rho_n) + k_o \frac{(\cos \phi_n)^2 - (\sin \phi_n)^2}{\rho_n} J_1(k_o \rho_n) \right] &= \\ &= k_o^2 \sin^2 \phi_n \cdot 1 + \left((\cos \phi_n)^2 - (\sin \phi_n)^2 \right) \frac{k_o}{\rho_n} \underbrace{\left(\frac{k_o \rho_n}{2} \right)}_{\text{small argument approximation}} \\ &= \frac{k_o^2}{2} (\sin^2 \phi_n + \cos^2 \phi_n) = \frac{k_o^2}{2} \end{aligned}$$

- For the constant term:

$$\circ \left(k_o^2 + \frac{\partial^2}{\partial x^2} \right) \frac{1}{4j} \left(-j \frac{4}{k_o^2} \right) = -1$$

- Finally,

$$\circ A_{n,n} = \frac{1}{\epsilon_{r_n} - 1} - \left(\frac{1}{4j} \frac{2\pi a_n k_o^2}{k_o} H_1^{(2)}(k_o a_n) - 1 \right) = \frac{\epsilon_{r_n}}{\epsilon_{r_n} - 1} - \frac{\pi k_o a_n}{4j} H_1^{(2)}(k_o a_n)$$

- By Duality,

$$\circ D_{n,n} = A_{n,n}$$

- For the off-diagonal blocks,

$$\begin{aligned} \frac{\partial^2}{\partial x \partial y} J_0^{(2)}(k_o \rho_n) &= \frac{(x - x_n)(y - y_n)}{\rho_n^3} [2J_1(k_o \rho_n) - k_o \rho_n J_0(k_o \rho_n)] \\ &= \frac{\cos \phi_n \sin \phi_n}{\rho_n} [2J_1(k_o \rho_n) - k_o \rho_n J_0(k_o \rho_n)] \end{aligned}$$

$$\circ \lim_{\rho_n \rightarrow 0} \frac{\cos \phi_n \sin \phi_n}{\rho_n} [2J_1(k_o \rho_n) - k_o \rho_n J_0(k_o \rho_n)] = \frac{\cos \phi_n \sin \phi_n}{\rho_n} \left[2 \frac{k_o \rho_n}{2} - k_o \rho_n \cdot 1 \right] = 0$$

- Therefore,
 - $B_{n,n} = C_{n,n} = 0$

The Impedance Matrix

- Summary:

$$j \frac{k_o}{\eta_o} \begin{bmatrix} \bar{e}_x \\ \bar{e}_y \end{bmatrix} = \begin{bmatrix} [A_{m,n}] & [B_{m,n}] \\ [C_{m,n}] & [D_{m,n}] \end{bmatrix} \begin{bmatrix} \bar{\alpha}_x \\ \bar{\alpha}_y \end{bmatrix}$$

$$A_{m,n} = -K_n \left[k_o^2 \frac{(y_m - y_n)^2}{\rho_{m,n}^2} H_0^{(2)}(k_o \rho_{m,n}) + k_o \frac{(x_m - x_n)^2 - (y_m - y_n)^2}{\rho_{m,n}^3} H_1^{(2)}(k_o \rho_{m,n}) \right], \quad m \neq n$$

$$D_{m,n} = -K_n \left[k_o^2 \frac{(x_m - x_n)^2}{\rho_{m,n}^2} H_0^{(2)}(k_o \rho_{m,n}) + k_o \frac{(y_m - y_n)^2 - (x_m - x_n)^2}{\rho_{m,n}^3} H_1^{(2)}(k_o \rho_{m,n}) \right], \quad m \neq n$$

$$C_{m,n} = B_{m,n} = -K_n k_o \frac{(y_m - y_n)(x_m - x_n)}{\rho_{m,n}^3} \left[2H_1^{(2)}(k_o \rho_{m,n}) - k_o \rho_n H_0^{(2)}(k_o \rho_{m,n}) \right], \quad m \neq n$$

$$\text{where } K_n = \frac{\pi a_n}{2 j k_o} J_1(k_o a_n)$$

$$A_{n,n} = D_{n,n} = \frac{\epsilon_{r_n}}{\epsilon_{r_n} - 1} - \frac{\pi k_o a_n}{4 j} H_1^{(2)}(k_o a_n), \quad C_{n,n} = B_{n,n} = 0$$

$$e_{x_m} = j k_o \frac{E_o}{\eta_o} \sin \phi^{inc} e^{j \bar{k}^{inc} \cdot \bar{r}_m}, \quad e_{y_m} = -j k_o \frac{E_o}{\eta_o} \cos \phi^{inc} e^{j \bar{k}^{inc} \cdot \bar{r}_m}$$

Far Field Approximation

- The scattered electric field is computed as:

$$E^{scat}(\vec{r}) = -jk_o\eta_o\vec{A}(\vec{r}) + \frac{\eta_o}{jk_o}\nabla\nabla\cdot\vec{A}(\vec{r})$$

- In the *far-field zone*

$$E^{scat}(\vec{r}) \approx -jk_o\eta_o\vec{A}(\vec{r})$$

- For power propagation in the radiation direction, we are interested in E_ϕ^{scat} :

- $E_\phi^{scat}(\vec{r}) \approx -jk_o\eta_o(-\sin\phi A_x(\vec{r}) + \cos\phi A_y(\vec{r}))$

- where: $A_{x,y}(\vec{r}) \approx \sum_{n=1}^N \alpha_{x,y,n} \frac{\pi a_n}{2jk_o} J_1(k_1 a_n) H_0^{(2)}(k_o \rho_n)$

- Far-field approximation:

$$\lim_{x \rightarrow \infty} H_0^{(2)}(k_o \rho_n) \approx \sqrt{\frac{2j}{\pi k_o \rho}} e^{-jk_o \rho} e^{jk_o \hat{\rho} \cdot \vec{r}_n}$$

- where, $k_o \hat{\rho} = \hat{x}k_o \cos\phi + \hat{y}k_o \sin\phi = \vec{k}$

- Thus, in the far field:

$$E_\phi^{scat,ff}(r, \phi) \approx \frac{\eta_o}{2} \sqrt{\frac{2j\pi}{k_o}} \frac{e^{-jk_o \rho}}{\sqrt{\rho}} \sum_{i=1}^N a_n J_1(k_o a_n) [\alpha_{x_i} \sin\phi - \alpha_{y_i} \cos\phi] e^{j\vec{k} \cdot \vec{r}_n}$$

Echo Width

- Recall, the echo width of the target is:

$$\sigma_{2d}(\phi) = \lim_{k_o \rho \rightarrow \infty} 2\pi\rho \frac{P^s}{P^{inc}} = \lim_{k_o \rho \rightarrow \infty} 2\pi\rho \frac{\hat{r} \cdot \vec{E}^{scat} \times \vec{H}^{scat*}}{|E^{inc}|^2 / \eta_o} = \lim_{k_o \rho \rightarrow \infty} 2\pi\rho \frac{|E_\phi^{scat}|^2}{|E^{inc}|^2}$$

- This leads to:

$$\sigma_{2d}(\phi) = \frac{\eta_o^2 \pi^2}{k_o E_o^2} \left| \sum_{i=1}^N a_n J_1(k_o a_n) [\alpha_{x_i} \sin \phi - \alpha_{y_i} \cos \phi] e^{j\vec{k} \cdot \vec{r}_n} \right|^2$$

- where E_o is the amplitude of the incident electric field.
- Or, since $H_o = E_o / \eta_o$

$$\sigma_{2d}(\phi) = \frac{\pi^2}{k_o H_o^2} \left| \sum_{i=1}^N a_n J_1(k_o a_n) [\alpha_{x_i} \sin \phi - \alpha_{y_i} \cos \phi] e^{j\vec{k} \cdot \vec{r}_n} \right|^2$$

MathCad Example

ka = 1 εr = 4 nx = 16 ny = 16 N = 861

Error = 0.035

