

1<sup>st</sup> Order Open-Loop System

$$H(s) = K/[(T)s+1]$$

Zero-order Hold:

$$G_{zoh}(s) = \frac{1 - e^{-sT_s}}{s} = \frac{T_s}{\frac{T_s}{2}s + 1} \cong \frac{1}{\frac{T_s}{2}s + 1}$$

2<sup>nd</sup> Order-Systems:

Quantizing: MSE = q<sup>2</sup>/12 where q is width of quantizing zone

$$\ddot{y} + 2\zeta\omega_n \dot{y} + \omega_n^2 y = \omega_n^2 w \Rightarrow \frac{Y(s)}{W(s)} = \frac{\omega_n^2}{s^2 + 2\zeta\omega_n s + \omega_n^2}$$

$$s_1 = -\zeta\omega_n \pm j\omega_n \sqrt{1 - \zeta^2}$$

$$t_s = 4 / \zeta\omega_n = -4 / \text{Re}(s)_{\max} \quad M_p = \exp\left(\frac{-\zeta\pi}{\sqrt{1 - \zeta^2}}\right) \quad t_p = \frac{\pi}{\omega_n \sqrt{1 - \zeta^2}}$$

Bilinear:  $H(z) = H(s) \Big|_{s = 2/T_s * (z-1)/(z+1)}$

$$y|_{u(t)} = 1 - e^{-\zeta\omega_n t} \left[ \cos(\omega_d t) + \left(\frac{\zeta}{\sqrt{1 - \zeta^2}}\right) \sin(\omega_d t) \right] \text{ where } \omega_d = \omega_n \sqrt{1 - \zeta^2} \quad \text{Mapping } z = e^{sT_s}$$

Type Number (q):  $G(s) = \frac{K(s/\omega_1 + 1)(s/\omega_2 + 1)\dots}{s^q(s/\omega_a + 1)(s/\omega_b + 1)\dots} \cong \frac{K}{s^q}$  at low frequencies

$$G_{PID} = K_i/s + K_p + K_d s$$

Error Analysis: (Valid for unity-feedback with closed loop system stable):

$$K_p = \lim_{s \rightarrow 0} G(s) \quad K_v = \lim_{s \rightarrow 0} sG(s) \quad \text{ess}|_{\text{step}} = 1 / (1 + K_p) \quad \text{ess}|_{\text{ramp}} = 1 / K_v$$

Delay:  $H_{\text{delay}}(s) = e^{-s\tau} = \frac{e^{-s\tau/2}}{e^{s\tau/2}} \cong \frac{1 - s\tau/2}{1 + s\tau/2}$  (Pade' approximation)

Eigenvalues:  $\det[sI - A] = 0$

State Variables:

$$J = \frac{1}{2} \int_0^\infty x^T Q x + w^T R w dt$$

State Model:  $\dot{x} = Ax + Bw, x(0)$   
 $y = Cx + Dw$

LQR:  $0 = A^T S(\infty) + S(\infty)A - S(\infty)BR^{-1}B^T S(\infty) + Q$

$$K(\infty) = R^{-1}B^T S(\infty)$$

Solution (zero-input):  $x(t) = e^{At} x(0) = P_1 e^{s_1 t} + P_2 e^{s_2 t} + \dots + P_n e^{s_n t}$

$$e^{At} = P e^{St} P^{-1} \text{ where } S = \begin{bmatrix} s_1 & 0 & 0 & \dots \\ 0 & s_2 & 0 & \dots \\ 0 & 0 & \ddots & 0 \\ \vdots & \vdots & 0 & s_n \end{bmatrix}, P = [P_1 \ P_2 \ \dots \ P_n], e^{St} = \begin{bmatrix} e^{s_1 t} & 0 & 0 & \dots \\ 0 & e^{s_2 t} & 0 & \dots \\ 0 & 0 & \ddots & 0 \\ \vdots & \vdots & 0 & e^{s_n t} \end{bmatrix}$$

Transfer Function:

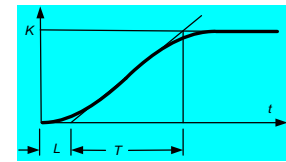
$$\frac{Y(s)}{W(s)} = C[sI - A]^{-1} B + D \quad \text{Regulator: } w = -Kx \text{ CL equation: } \dot{x} = (A - BK)x \text{ Design } K \text{ to set the eigenvalues}$$

$$M = [B \ AB \ A^2 B \ \dots \ A^{n-1} B] \quad \text{rank}(M) = \# \text{ of controllable eigenvalues}$$

Sensitivity:  $S_y^x = \frac{\partial x}{\partial y} \frac{Y}{X} \Big|_{x_{nom}, y_{nom}}$  Disturbance Rejection:  $\frac{\Delta Y(s)}{\Delta D(s)} = \frac{G_{zoh}G(s)}{1 + G_c G_{zoh}G(s)} \cong \frac{1}{G_c(s)}$  if  $G_c G_{zoh}G(s) \gg 1$

Observer Equation:  $\hat{x} = A\hat{x} + Bw + K_o(y - C\hat{x})$

Ziegler-Nichols Method I



$$G_c(s) = K_p \left( 1 + K_D s + \frac{1}{K_I s} \right)$$

Controller	$\bar{K}_P$	$\bar{K}_I$	$\bar{K}_D$
P	$T/L$	$\infty$	0
PI	$0.9T/L$	$L/0.3$	0
PID	$1.2T/L$	$2L$	$0.5L$

Method II – tune  $K_0$  until CL is

marginally stable. Record period P

Controller	$\bar{K}_P$	$\bar{K}_I$	$\bar{K}_D$
P	$0.5K_0$	$\infty$	0
PI	$0.45K_0$	$P/1.2$	0
PID	$0.6K_0$	$0.5P$	$0.125P$